

Broad- and Narrow-Line Terahertz Filtering in Frequency-Selective Surfaces Patterned on Thin Low-Loss Polymer Substrates

Antonio Ferraro, Dimitrios C. Zografopoulos, Roberto Caputo, and Romeo Beccherelli

Abstract—A new class of frequency-selective surface filters (FSS) for terahertz (THz) applications is proposed and investigated both numerically and experimentally. A periodic FSS array of cross-shaped apertures is patterned on aluminum, deposited on thin foils of the low-loss cyclo-olefin polymer Zeonor. Apart from the fundamental filtering response of the FSS elements, we also observe very narrow-linewidth peaks with high transmittance, associated with guided-mode resonances in the dielectric substrate. The effect of the filter's geometrical parameters on its performance is systematically studied via finite-element simulation and confirmed by time-domain spectroscopy characterization of the fabricated samples. Finally, thanks to the flexibility of the employed substrates, THz-FSS filters are also characterized in bent configuration, revealing a robust response in terms of the fundamental FSS passband filter and a high sensitivity of the GMR peaks. These features can be exploited in the design of novel THz filters or sensors.

Index Terms—Terahertz photonics, frequency-selective surfaces, terahertz filters, grating mode resonances, flexible devices.

I. INTRODUCTION

THE terahertz (THz) frequency range has been under intense investigation due to its numerous applications from fundamental to applied science, among which secure short-range communications, life-science diagnostics, defense and security [1]–[5]. Furthermore, this technological interest towards THz science is constantly growing, driven by recent advances in the development of novel and relatively low-cost THz sources with improved performance. In this context, the design of new components capable of manipulating the amplitude, phase, or

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polarization, of THz radiation is of paramount importance, and several research groups have been working on the development of such components, e.g. polarizers [6], phase shifters [7], [8], electromagnetic absorbers [9], [10], or tunable filters [11], [12].

Photoconductive antennas and non-linear organic or inorganic crystals illuminated by fs laser pulses represent perhaps the most widely used THz sources [13], typically employed in terahertz time domain spectroscopy (THz-TDS) setups. However, their radiation is broadband, spanning a range of a few THz. In view of that, selective filters that allow for the bandpass transmission around a specific frequency are well needed. Such filters can be used also in other fields, for instance astronomy, telecommunications, imaging, detection, or radar science [14], [15], and have long attracted the attention of many researchers. The most common typology of bandpass filters in the THz frequency range is based on resonant frequency selective surfaces (FSS), already known since 1983 [16], where a polarization-independent square periodic array of cross-shaped apertures was patterned on a free-standing nickel foil. Later, Porterfield *et al.* [17] applied the same geometry using copper; these two seminal works established the main design rules and explain how the performance of the filters depends on their geometry.

Nevertheless, free-standing FSS filters or, in general, THz components are fragile and need a mechanical support [18]–[20], which raises their cost and may render both their fabrication and use problematic. To alleviate this problem, various attempts have been done to fabricate FSS on dielectric substrates. The choice of the substrate material is critical as it has to be very low-loss at THz frequencies so as not to compromise the filter's transmittance. One approach is the patterning of the FSS or any metasurface, on a few micron-thick polyimide films, either via photolithography [21], [22] or other techniques, such as the more time and cost-consuming UV-laser direct writing [23], [24]. This results in membrane-like flexible samples, which still need some kind of mechanical support. The use of thick substrates introduces rigidity but this may come at the expense of higher losses and hence lower transmittance, as in FSS filters fabricated on 1-mm thick high-density polyethylene substrates [25], [26]. In [27] a FSS filter patterned on both sides of a costly 525- μm high-resistivity silicon substrate, which presents very good out-of-band rejection, although still accompanied by moderate peak transmittance. Other substrate solutions include the use of 100- μm polyethylene terephthalate

[28], [29] or naphthalate [30], which were employed in a different context, namely the design and fabrication of multi-layer stacks of split-ring resonator FSS and metamaterials in the THz spectrum [31]–[34].

In this work, we present a numerical and experimental study of aluminum-based, cross-shaped aperture FSS filters fabricated on thin substrates made of Zeonor, a low-cost cyclo-olefin polymer that shows very low losses at THz frequencies. Apart from the well-known broad-line filtering response that stems from the metallic mesh FSS, we observe a series of Fano-like asymmetric narrow-line transmission peaks at higher frequencies, inside the diffraction regime.

It is well known that when the operating wavelength is smaller than the FSS lattice pitch part of the THz radiation is diffracted, which may lead to various interesting phenomena, such as diffractive coupling between adjacent FSS elements [35]–[37]. However, in the case of the here investigated metallic THz filters, we observe guided-mode resonances (GMR) that occur at resonant frequencies where the first-order diffracted waves are phased-matched and thus coupled to modes guided in the dielectric substrate. These GMR are responsible for the narrow-line peaks and their influence on the filter's performance is thoroughly investigated.

The experimental observation of GMR at THz frequencies has been only very recently reported [38] at the frequency of approximately 7 THz. Although demonstrating the proof-of-principle, these observed GMR peaks exhibited low transmittance and broad lines owing to the use of a lossy substrate. Theoretically, it has been proposed that GMR can lead to very narrow-linewidth resonances that manifest in the spectral response of THz metamaterials [39]. In this work, we have experimentally measured THz resonances with transmittance well above 50% and full-width half-maximum (FWHM) even below 1% of the resonant frequency.

The FSS filters are fabricated using standard lithography processes, which can be scaled up to mass production using low-cost large area electronics and roll-to-roll processes. Very good agreement is observed between experimental THz-TDS measurements and finite-element numerical simulations. The resulting samples are both mechanically stable and conformable. The latter property allowed for the characterization of the filters in a bent configuration, which revealed two distinct behaviors, the robustness of the fundamental FSS filter response and the suppression of the GMR peaks. Finally, a discussion on possible applications of this novel class of FSS-THz filters in low-cost and flexible THz devices is provided.

II. NUMERICAL ANALYSIS

The schematic layout of the proposed THz-FSS filters is shown in Fig. 1. The periodic metallic FSS square lattice is characterized by the pitch P and the cross-shaped apertures are defined by the cross-arm length W and width w . The metallic screen is made of aluminum and it is supported by a thin foil of the cyclo-olefin polymer Zeonor of thickness d . The Al film thickness is 200 nm, which is thicker than the Al skin depth in the investigated frequency range, namely 150 and 58 nm at

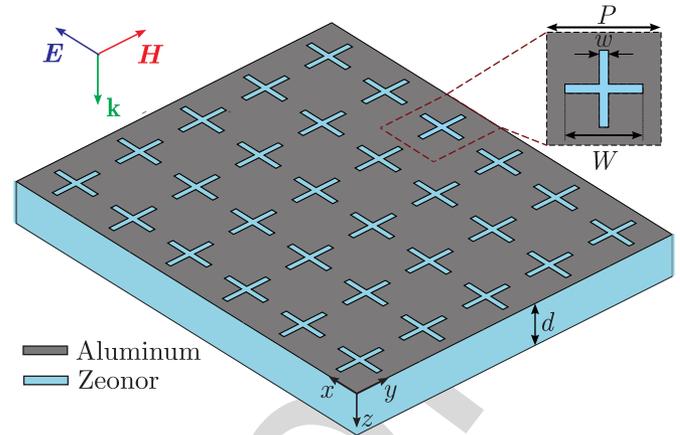


Fig. 1. Schematic layout of the investigated FSS terahertz filters. The square lattice has a pitch P and the length and width of the cross-shaped apertures are W and w , respectively. The FSS structure is patterned on a 200-nm-thick aluminum layer deposited on a Zeonor foil of subwavelength thickness d .

0.3 and 2 THz, respectively. At the same time, it is thin enough in order to avoid unnecessary stressing of the polymer during the fabrication process. The thickness of the Zeonor layer is in principle an independent variable, however in this study we focus on three values, $d = 40, 100, \text{ and } 188 \mu\text{m}$, which correspond to available films for fabrication, as it will be discussed in Section III.

The numerical simulations of the THz filters was conducted via the frequency-domain finite element method (FEM), which was implemented in the commercial software COMSOL Multiphysics. A unit cell of the periodic array was simulated by properly imposing periodic boundary conditions at the $x-z$ and $y-z$ planes. The structure was excited with an x -polarized plane wave propagating along the z -axis as in Fig. 1. The transmittance of the zero-order diffracted mode, i.e. the excited planewave, was measured at the exit of the filter, below the polymer film substrate, and normalized to the power carried by the excitation planewave. Aluminum was modeled as a Drude medium [40] and Zeonor as a dielectric with a refractive index equal to $n_z = 1.52 - j0.001$, as it has been demonstrated that it exhibits very low dispersion in the frequency range under investigation [41]–[43]. This polymeric material was selected for its excellent THz properties, namely very low-losses, high mechanical flexibility, heat resistance, and negligible birefringence. In fact, it was observed via numerical simulations that, in the context of the proposed THz-FSS filters and for the substrate thicknesses here reported, the effect of Zeonor's dielectric losses was negligible.

Fig. 2(a) provides a reference result on the transmittance of a free-standing THz filter, i.e. in the absence of the polymer substrate, for $P = 160 \mu\text{m}$, $W = 110 \mu\text{m}$, and $w = 10 \mu\text{m}$. A Lorentzian-shaped filter is observed with peak transmittance $T = 0.88$, resonant frequency $f_0 = 1.293 \text{ THz}$, and a FWHM of 200 GHz, i.e. $\sim 15\%$ of f_0 . The filtering effect stems from the response of THz wave transmission through the cross-shaped apertures owing to the resonance of the fundamental mode in the cross arm slots, which is maximized at the frequency f_0 ,

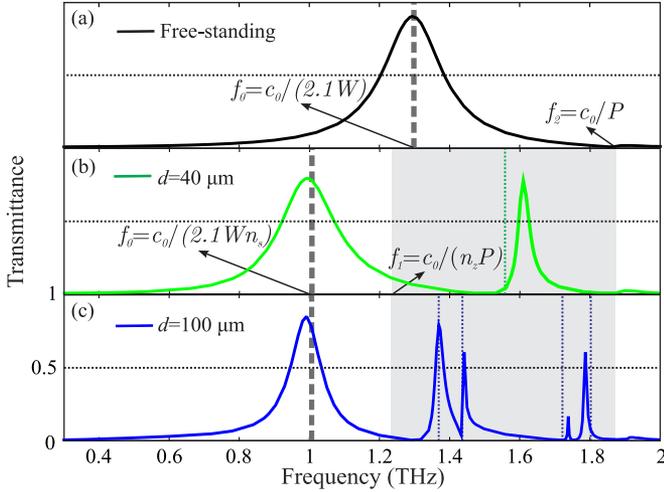


Fig. 2. (a) Power transmittance of the zero-order diffracted mode, numerically calculated for a free-standing FSS filter with $P = 160 \mu\text{m}$, $W = 110 \mu\text{m}$, and $w = 10 \mu\text{m}$. The dashed line corresponds to the frequency f_0 predicted by the approximative formula for resonant dipoles, whereas f_2 denotes the onset of the first diffractive order, associated with Wood's anomaly that leads to zero transmittance at the frequency f_2 . (b,c) Transmittance of the filter for a substrate thickness $d = 40 \mu\text{m}$ and $d = 100 \mu\text{m}$, respectively, where other parameters as in (a). The shaded region denotes the frequency interval where guided-mode resonance can manifest and the dotted lines mark the resonant frequencies predicted by GMR theory.

169 due to impedance matching [44]. When $w/(P - W) < 1$, the
 170 resonant frequency can be well approximated by the formula for
 171 resonant dipoles $f_0 = c_0/(2.1W)$ [16], marked as a dashed line
 172 in Fig. 2 and thereafter. As in other similar filters reported in the
 173 literature, this resonance only marginally depends on the pitch
 174 P , since it is not diffractive in nature, and in this work we denote
 175 it as FSS resonance (FSSR). At the frequency $f_2 = c_0/P$ that
 176 marks the threshold at which the first diffractive order appears,
 177 c_0 being the speed of light in free-space, the Wood's anomaly
 178 associated with zero transmittance is observed [21], [45].

179 Fig. 2(b) and (c) investigate the same FSS structure, albeit in
 180 the presence of the Zeonor substrate, with a thickness of 40 and
 181 100 μm , respectively. Compared to the free-standing reference
 182 case of Fig. 2(a), two major differences are observed. First, the
 183 FSSR is shifted towards lower frequencies by a factor

$$n_s = \sqrt{\frac{(n_{zr}^2 + 1)}{2}}, \quad (1)$$

184 where n_{zr} is the real part of n_z , and n_s is the index correspond-
 185 ing to the average permittivity of materials on the two sides
 186 of the FSS, namely air and Zeonor. Equation (1) is valid for
 187 substrate thicknesses higher than one tenth of the THz wave-
 188 length [21], which is the case for the considered values of d .
 189 In general, in the presence of a substrate or superstratum the
 190 THz wave has a shorter wavelength in the dielectric medium,
 191 the cross dimensions become electromagnetically larger and
 192 the FSSR frequency decreases. Second, apart from the FSSR,
 193 other resonances are observed, whose number and position de-
 194 pends on the polymer thickness. These resonances stem from
 195 the coupling of waves diffracted on the periodic FSS screen to

196 propagating modes in the substrate, which can be thought of
 197 as a dielectric slab waveguide, a phenomenon known as GMR.
 198 Grating filters based in GMR have been long known in the field
 199 of optics and photonics [46] and recently it has been shown that
 200 bandpass GMR filters can also be designed at THz frequencies,
 201 where the role of the anti-reflecting surface [47] can be played
 202 by the metallic FSS layer [38].

203 According to GMR theory, first-order resonances for normal
 204 incidence can be observed in the interval

$$\frac{c_0}{n_{zr}P} \leq f_r \leq \frac{c_0}{P}, \quad (2)$$

205 at those resonant frequencies f_r that satisfy $n_i = c_0/(f_r P)$,
 206 where n_i is the effective index of a mode guided in the substrate
 207 slab waveguide. In Fig. 2(b) and (c) we have annotated with
 208 grey shading the spectral window where GMR can occur. In-
 209 side these regions, we have calculated a set of GMR frequencies,
 210 marked as dotted lines, according to the following steps: first, the
 211 resonant frequencies f_r calculated by the FEM simulations are
 212 identified, i.e. the transmission maxima in the gray-shaded areas
 213 of the calculated spectra. Then, for each slab thickness the effec-
 214 tive modal indices at f_r are calculated using a freely available
 215 electromagnetic mode solver for 1-D dielectric multilayer slab
 216 waveguides [48]. Among the resulting modal indices $n_i(f_r)$, the
 217 frequencies $c_0/(n_i P)$ are calculated and the one closely match-
 218 ing f_r is marked, with each resonant frequency associated with
 219 a different slab mode. Better agreement is achieved for higher
 220 values of d and for modes closer to the limit $f_1 = c_0/(n_{zr}P)$. In
 221 both cases, these modes show higher confinement thanks to ei-
 222 ther the higher slab thickness or the higher effective modal index
 223 and hence lower modal order. The discrepancy between GMR
 224 theory and FEM simulations is attributed to the presence of the
 225 reflecting metallic FSS screen, which introduces a perturbation
 226 in the geometry of the slab waveguide.

227 It is clear that the positions of the GMR frequencies depend
 228 strongly on d and P , as these parameters control the matching
 229 condition between the wave vector of the diffracted orders and
 230 the propagating substrate modes. This strong dependence is not
 231 to be expected as far as the exact geometry of the cross-shaped
 232 apertures is concerned. In order to further elucidate this point,
 233 we have calculated the transmittance of a series of filters with
 234 fixed $d = 100 \mu\text{m}$, $P = 160 \mu\text{m}$, $w = 10 \mu\text{m}$, for various val-
 235 ues of the cross-arm length W . The results reported in Fig. 3
 236 demonstrate that the GMR are only slightly affected by the
 237 variation of W , as summarized in the results of Table I, while
 238 the opposite stands for the FSSR, which depends on W via
 239 $f_0 = c_0/(2.1Wn_s)$. Fig. 4 investigates a complementary sce-
 240 nario, namely the variation of the pitch P for an FSS with fixed
 241 $W = 110 \mu\text{m}$, $w = 10 \mu\text{m}$, for the three available thicknesses
 242 of the Zeonor film thickness. In this case it is the FSS resonant
 243 frequency f_0 that remains approximately at the same position,
 244 while those of the GMR shift towards lower frequencies for
 245 higher pitch values, given the condition described by (2). More-
 246 over, as d gets higher, a larger number of GMR is supported
 247 that corresponds to a higher number of modes guided in the slab
 248 waveguide, whose positions are well resolved by the numerical
 249 simulations described above.

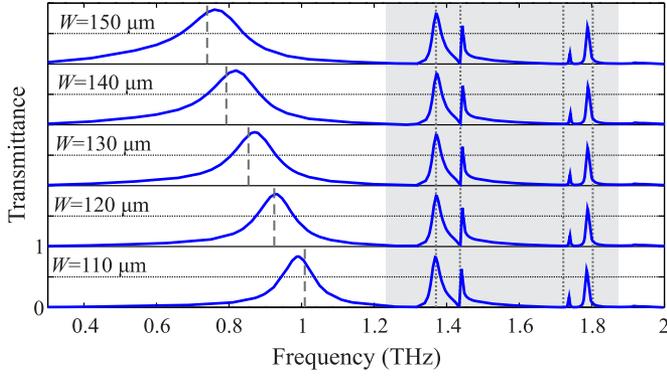


Fig. 3. Transmittance of the FSS filter for $d = 100 \mu\text{m}$, $P = 160 \mu\text{m}$, $w = 10 \mu\text{m}$, for various values of the cross-arm length W . The main resonant frequency f_0 shifts towards lower frequencies for higher W , whereas the GMRs remain unaffected. The spectral position of the GMRs is well approximated by calculating the frequencies $f_i = c_0 / (n_{\text{eff},i} P)$, marked as dashed lines, where n_i are the effective indices of the modes supported by the single Zeonor slab waveguide.

TABLE I
SIMULATED GUIDED-MODE RESONANT FREQUENCIES (IN THZ)
FOR THE FILTERS STUDIED IN FIG. 3

W (μm)	$f_r^{1,\text{sim}}$	$f_r^{2,\text{sim}}$	$f_r^{3,\text{sim}}$	$f_r^{4,\text{sim}}$
150	1.371	1.443	1.743	1.797
140	1.372	1.445	1.749	1.803
130	1.373	1.448	1.753	1.809
120	1.375	1.451	1.753	1.815
110	1.379	1.453	1.753	1.817

The resonances predicted via GMR theory (dashed lines) occur at $f_r^1 = 1.37 \text{ THz}$, $f_r^2 = 1.438 \text{ THz}$, $f_r^3 = 1.733 \text{ THz}$, and $f_r^4 = 1.803 \text{ THz}$, and the corresponding slab modal indices [48] are $n_1 = 1.368$, $n_2 = 1.302$, $n_3 = 1.081$, and $n_4 = 1.038$.

250 Apart from the interesting underlying physics, the GMR fil-
251 ters show also great potential in view of THz applications that
252 need narrow-line selective filtering. In this respect, one main
253 drawback of the traditional FSSR filters is that their linewidth,
254 defined as FWHM/f_0 , is in the range $5\% \sim 20\%$, according to
255 the various designs [38]. Achieving more narrow filters is pos-
256 sible by reducing the aperture's dimensions, particularly w , or
257 by stacking more than one FSS filters, although this comes to
258 the expense of significantly reduced transmittance. On the con-
259 trary GMR-based THz filters can have linewidths lower than 1%
260 [38], without compromising the filter's transmittance, which is
261 validated both numerically and experimentally in this work.

262 III. EXPERIMENTAL DEMONSTRATION

263 We have experimentally investigated the properties of the
264 proposed THz filters by fabricating samples with different pe-
265 riod and cross-arm length values on low-loss flexible 40, 100,
266 and $188\text{-}\mu\text{m}$ -thick Zeonor foils using standard photolithogra-
267 phy techniques. First, an aluminum layer of 200 nm thickness
268 was thermally evaporated on Zeonor foils. Subsequently, a film
269 of the positive photoresist S1813 from Shipley was deposited

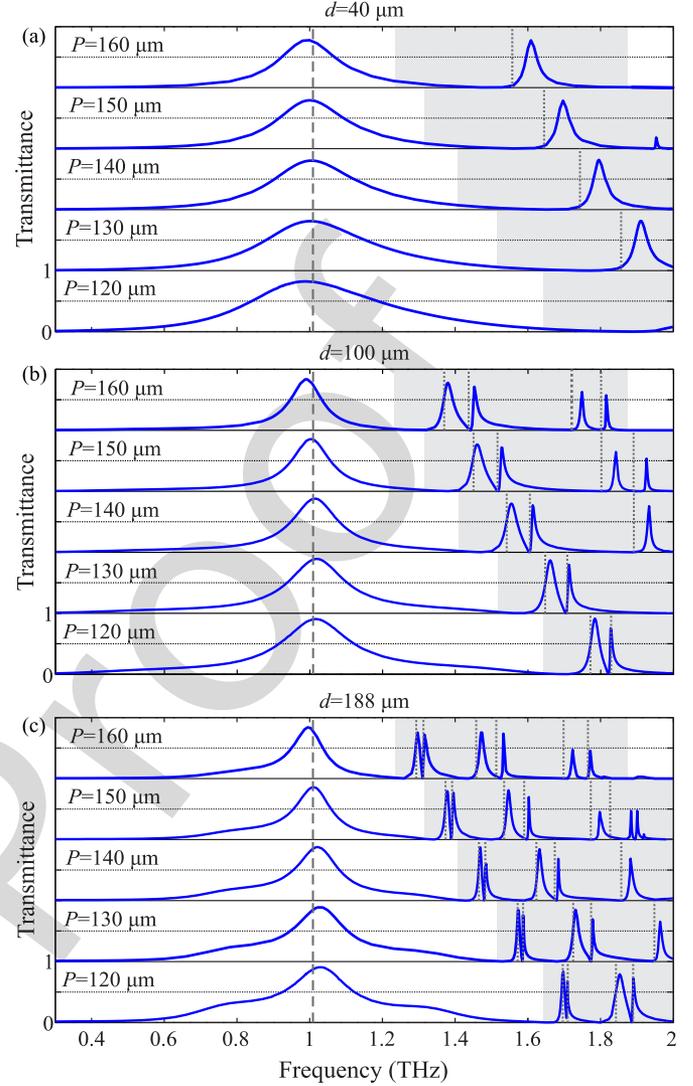


Fig. 4. Transmittance of the FSS filter numerically calculated for $W = 110 \mu\text{m}$, $w = 10 \mu\text{m}$, various values of the lattice pitch P and for three commercially available thicknesses of the Zeonor substrate, namely (a) $d = 40 \mu\text{m}$, (b) $d = 100 \mu\text{m}$, and (c) $d = 188 \mu\text{m}$. The main resonant frequency f_0 remains fixed at $f_0 \simeq c_0 / (2.1 P n_s)$, whereas the GMRs shift as a function of the lattice pitch and the substrate thickness.

270 by spin-coating at 3000 rpm for 30 seconds and then cured at
271 $115 \text{ }^\circ\text{C}$ for 2 minutes . The resulting thickness of the photoresist
272 layer was $1.3 \pm 0.1 \mu\text{m}$. Photolithography was carried out on
273 the metalized surface using a Karl Suss MA150 mask aligner
274 with a wavelength of 365 nm and intensity of $60 \text{ mW}/\text{cm}^2$. The
275 samples were immersed in the developer MF319 for 50 seconds ,
276 rinsed with deionized water, dried with nitrogen and cured at
277 $120 \text{ }^\circ\text{C}$ for 5 minutes . Then, the exposed aluminum was wet-
278 etched in $\text{H}_3 \text{ PO}_4 : \text{H}_2 \text{ O} : \text{CH}_3 \text{ COOH} : \text{HNO}_3 = 16:2:1:1$ and the
279 residual photoresist was removed with acetone. The filters were
280 cut in samples of $2 \text{ cm} \times 2 \text{ cm}$.

281 The transmission properties of the fabricated THz filters were
282 investigated by means of THz time domain spectroscopy using
283 a Menlo Systems TERA K15 THz-TDS all fiber-coupled spec-
284 trometer in transmission mode using collimated and polarized

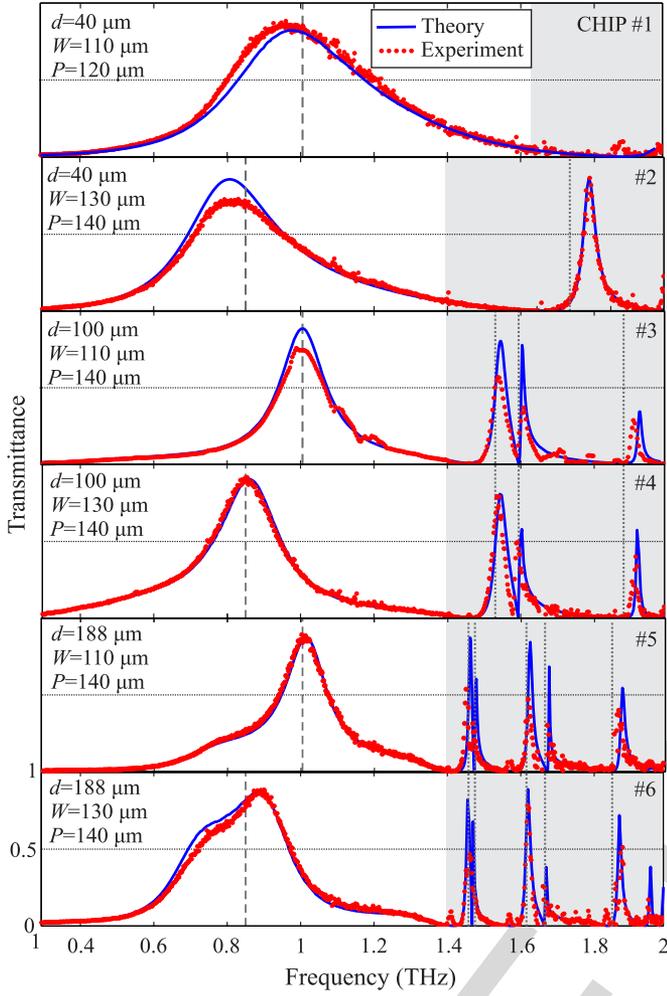


Fig. 5. Direct comparison of the FSS filter's transmittance between numerical FEM simulations and experimental TDS measurements for a series of fabricated samples with different geometrical parameters.

radiation. The power transmittance measured for each sample was normalized to that of the reference signal, i.e. in the absence of the sample. The spot-size of the collimated beam was approximately 10 mm in diameter and a time scan of 400 ps was employed for a spectral resolution of 2.5 GHz. The measurements were done in an atmosphere purged with nitrogen to prevent the absorption of THz radiation from water vapor in the air.

Fig. 5 shows a direct comparison between the numerically simulated transmittance of the THz filters, calculated by means of the finite-element method, and the experimental TDS measurements for six different chips, two for each one of the available Zeonor foil substrates. The experimental results reproduce very well the numerical simulations, in terms of both the position and the lineshape of the various transmission peaks. For $d = 40 \mu\text{m}$ and $P = 140 \mu\text{m}$ (Chip #2) two clearly separated resonances are observed, the FSSR at 0.8 THz and a single GMR at 1.8 THz. It is experimentally verified that, owing to the different underlying physical mechanisms, these two resonances show very different linewidths: the FWHM measured for

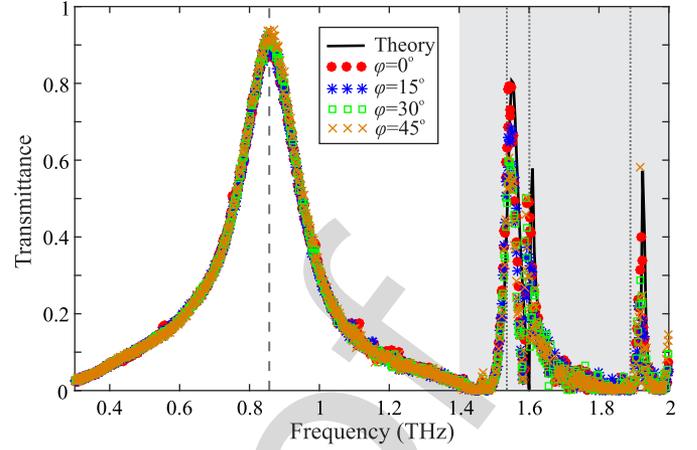


Fig. 6. TDS measurements of Chip #4, as in Fig. 5, for different angles φ of the sample's rotation in the $x-y$ plane, demonstrating polarization-insensitive operation.

the FSSR and GMR is 40% and only 2.3% of the resonant frequency, respectively. For $d = 100 \mu\text{m}$ and $188 \mu\text{m}$, the existence of closely spaced GMR peaks, stemming from the excitation of more modes in the slab waveguide substrate, leads to asymmetric Fano-like linewidths [49], with high transmittance and even more narrow linewidths. For instance, Chip #4 is characterized by three GMR at 1.546, 1.597, 1.921 THz with linewidths FWHM/f_0 equal to 2.5%, 1.3%, and 1%, respectively, while the high-transmittance resonances at 1.63 and 1.88 THz for Chip #6 exhibit corresponding linewidths of 1% and 0.7%.

It is observed that in the case of some GMR the experimentally measured transmittance peak values are somewhat lower than the numerically calculated prediction. This is attributed to three factors: a) the very narrow linewidths of such resonances, particularly for $d = 188 \mu\text{m}$, which are comparable with the TDS resolution, b) minor defects in fabrication or the planarity of the samples, c) non-ideal collimation/ residual divergence of the spot, and d) the finite dimensions of the sample and THz spot, with a diameter of a few tens of wavelength in size, in contrast to the infinite periodic array assumed in the simulations. The latter, in particular, is very relevant for GMR, since these are numerically simulated as the result of constructive interference of the excited waveguide modes and the diffractive waves on an infinite periodic FSS metallic screen, while the measurements are conducted over a finite truncated lattice. Nevertheless, apart from these small discrepancies, it is overall demonstrated that the fabricated FSS-THz filter can achieve both broad- and narrow-band filtering, depending on the selection of the geometrical parameters.

An important trait of the proposed THz filters is that they are polarization-independent, owing to the square FSS lattice and the symmetry of the cross-shaped apertures. This has been experimentally verified by rotating the fabricated samples in the $x-y$ plane, i.e. perpendicular to the propagation direction of the x -polarized THz wave, and recording the measured TDS spectra. Fig. 6 shows a set of results obtained for the filter characterized by $P = 140 \mu\text{m}$, $W = 110 \mu\text{m}$, $w = 10 \mu\text{m}$,

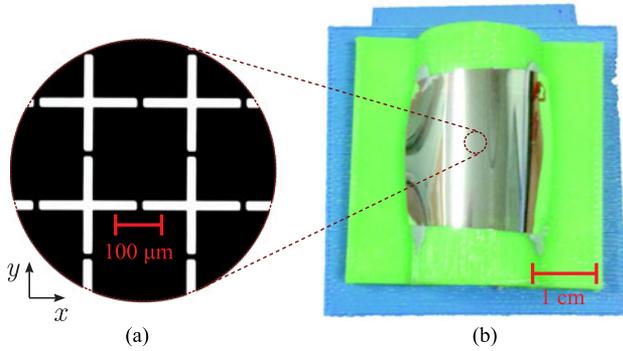


Fig. 7. (a) Micrograph taken under the microscope in transmission mode of a fabricated sample with $P = 140 \mu\text{m}$, $W = 130 \mu\text{m}$, and $w = 10 \mu\text{m}$ on a Zeonor foil with $d = 100 \mu\text{m}$. (b) The same sample bent at a radius of 1 cm and fixed on a properly assembled mount in order to characterize its properties as a conformal THz filter.

and $d = 100 \mu\text{m}$ (Chip #4 with reference to Fig. 5), where the angle φ denotes the rotation angle of the sample, measured from the x -axis. The spectra recorded for $\varphi = 0^\circ$, 15° , 30° , and 45° overlap, thus demonstrating the polarization-independent response of the THz filter.

Among the appealing properties of the employed thin polymer films are their flexibility and ability to easily conform to curved surfaces [6]. In this work, we have experimentally investigated the transmission properties of Chip #4, when bent down to a curvature with radius 1 cm. The inset of Fig. 7(a) shows a micrograph of the fabricated filter taken under optical microscope in transmission mode with a 20x microscope objective, where the black parts are aluminum, while the transparent foil appears white. Fig. 7(b) shows the bent sample mounted on a properly designed frame, so that it is placed in the THz beam path of the TDS setup. It is remarked that, after processing, the filter does not show any buckling and maintains its mechanical and electromagnetic properties after bending it several times.

Fig. 8 shows the experimental characterization of the bent THz filter, where the numerical and experimental results for the flat configuration are also reported for comparison. In the experimental characterization, the incoming THz wave was polarized along the x -axis, as defined in Fig. 7. It is evident that the bent filter retains its filtering property as far as the FSSR is concerned, while all remaining peaks that stem from GMR excitation are no longer observed. These interesting features can be explained by taking into account the physical origin of the filter's resonances. The FSSR involves the excitation of a localized mode inside the cross-shaped aperture, which depends on the aperture's dimensions and is not diffractive in nature, hence the very weak dependence of its central frequency on the lattice pitch, as demonstrated in Fig. 4. Also, the presence of the substrate induces a shift of the resonant frequency by the factor n_s , but does not otherwise affect the transmission mechanism of the filter.

On the contrary, the GMR are excited due to the coupling of first-order diffracted waves into the substrate slab modes. This coupling is strongly dependent on both the lattice pitch and the polymer film's thickness. When the sample is bent,

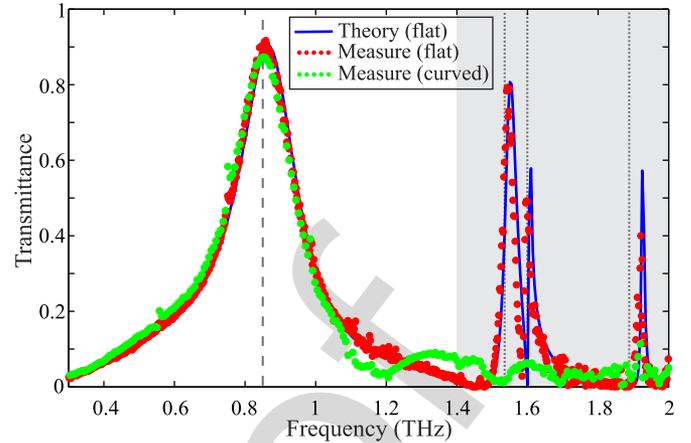


Fig. 8. Comparison of the experimentally measured transmittance of the FSS filter shown in Fig. 7 between the flat and bent configuration. The numerical simulations for the flat case are also shown for reference. The main FSS resonance remains unaffected, demonstrating the capability of the fabricated samples to operate as flexible and conformal thin film THz filters. The GMR peaks in the bent configuration are suppressed.

impinging THz plane wave does not sample the same effective pitch across the filter's surface and the substrate is no longer a flat slab dielectric waveguide. This leads to the suppression of GMR peaks, as evidenced in the results of Fig. 8. Therefore, apart from the filtering properties already demonstrated in Fig. 5 for the flat configuration, the proposed THz filters show very interesting properties also when bent, which can be readily exploited in various applications. For instance, flexible filters based on the FSSR can be designed for use as thin conformal layers on curved surfaces or components in THz setups. Moreover, the sensitivity of the GMR transmittance and central frequencies on the bending radius can provide the basis for the development of sensor devices, for instance curvature sensors or components for the measurement of the thickness and/or refractive index of thin dielectric layers at THz frequencies by placing the FSS-GMR filters on top of the sample and measuring the shift of the GMR frequencies.

IV. CONCLUSION

In brief, we have numerically and experimentally investigated a new class of THz filters based on the patterning of metallic cross-shaped FSS on thin films of the low-loss cyclo-olefin polymer Zeonor. By properly adjusting the geometrical parameters of the device both broad- and narrowline filters can be designed, the first stemming from the transmittance of THz waves through FSS cross-shaped apertures, while the latter from the excitation of guided-mode resonances in the polymer substrate. Not observed before in this kind of FSS structures, the GMR filters show extremely narrow linewidths with high transmittance. The FSS filters are shown robust to the bending of the flexible Zeonor films, thus paving the way for conformal THz filters integrated on curved surfaces. On the contrary, the diffractive nature of GMR renders them sensitive to deformations or changes of

413 substrate's properties, properties that could be exploited in the
414 design of sensors working at THz frequencies.

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IEEE PROOF

Broad- and Narrow-Line Terahertz Filtering in Frequency-Selective Surfaces Patterned on Thin Low-Loss Polymer Substrates

Antonio Ferraro, Dimitrios C. Zografopoulos, Roberto Caputo, and Romeo Beccherelli

Abstract—A new class of frequency-selective surface filters (FSS) for terahertz (THz) applications is proposed and investigated both numerically and experimentally. A periodic FSS array of cross-shaped apertures is patterned on aluminum, deposited on thin foils of the low-loss cyclo-olefin polymer Zeonor. Apart from the fundamental filtering response of the FSS elements, we also observe very narrow-linewidth peaks with high transmittance, associated with guided-mode resonances in the dielectric substrate. The effect of the filter's geometrical parameters on its performance is systematically studied via finite-element simulation and confirmed by time-domain spectroscopy characterization of the fabricated samples. Finally, thanks to the flexibility of the employed substrates, THz-FSS filters are also characterized in bent configuration, revealing a robust response in terms of the fundamental FSS passband filter and a high sensitivity of the GMR peaks. These features can be exploited in the design of novel THz filters or sensors.

Index Terms—Terahertz photonics, frequency-selective surfaces, terahertz filters, grating mode resonances, flexible devices.

I. INTRODUCTION

THE terahertz (THz) frequency range has been under intense investigation due to its numerous applications from fundamental to applied science, among which secure short-range communications, life-science diagnostics, defense and security [1]–[5]. Furthermore, this technological interest towards THz science is constantly growing, driven by recent advances in the development of novel and relatively low-cost THz sources with improved performance. In this context, the design of new components capable of manipulating the amplitude, phase, or

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polarization, of THz radiation is of paramount importance, and several research groups have been working on the development of such components, e.g. polarizers [6], phase shifters [7], [8], electromagnetic absorbers [9], [10], or tunable filters [11], [12].

Photoconductive antennas and non-linear organic or inorganic crystals illuminated by fs laser pulses represent perhaps the most widely used THz sources [13], typically employed in terahertz time domain spectroscopy (THz-TDS) setups. However, their radiation is broadband, spanning a range of a few THz. In view of that, selective filters that allow for the bandpass transmission around a specific frequency are well needed. Such filters can be used also in other fields, for instance astronomy, telecommunications, imaging, detection, or radar science [14], [15], and have long attracted the attention of many researchers. The most common typology of bandpass filters in the THz frequency range is based on resonant frequency selective surfaces (FSS), already known since 1983 [16], where a polarization-independent square periodic array of cross-shaped apertures was patterned on a free-standing nickel foil. Later, Porterfield *et al.* [17] applied the same geometry using copper; these two seminal works established the main design rules and explain how the performance of the filters depends on their geometry.

Nevertheless, free-standing FSS filters or, in general, THz components are fragile and need a mechanical support [18]–[20], which raises their cost and may render both their fabrication and use problematic. To alleviate this problem, various attempts have been done to fabricate FSS on dielectric substrates. The choice of the substrate material is critical as it has to be very low-loss at THz frequencies so as not to compromise the filter's transmittance. One approach is the patterning of the FSS or any metasurface, on a few micron-thick polyimide films, either via photolithography [21], [22] or other techniques, such as the more time and cost-consuming UV-laser direct writing [23], [24]. This results in membrane-like flexible samples, which still need some kind of mechanical support. The use of thick substrates introduces rigidity but this may come at the expense of higher losses and hence lower transmittance, as in FSS filters fabricated on 1-mm thick high-density polyethylene substrates [25], [26]. In [27] a FSS filter patterned on both sides of a costly 525- μm high-resistivity silicon substrate, which presents very good out-of-band rejection, although still accompanied by moderate peak transmittance. Other substrate solutions include the use of 100- μm polyethylene terephthalate

[28], [29] or naphthalate [30], which were employed in a different context, namely the design and fabrication of multi-layer stacks of split-ring resonator FSS and metamaterials in the THz spectrum [31]–[34].

In this work, we present a numerical and experimental study of aluminum-based, cross-shaped aperture FSS filters fabricated on thin substrates made of Zeonor, a low-cost cyclo-olefin polymer that shows very low losses at THz frequencies. Apart from the well-known broad-line filtering response that stems from the metallic mesh FSS, we observe a series of Fano-like asymmetric narrow-line transmission peaks at higher frequencies, inside the diffraction regime.

It is well known that when the operating wavelength is smaller than the FSS lattice pitch part of the THz radiation is diffracted, which may lead to various interesting phenomena, such as diffractive coupling between adjacent FSS elements [35]–[37]. However, in the case of the here investigated metallic THz filters, we observe guided-mode resonances (GMR) that occur at resonant frequencies where the first-order diffracted waves are phased-matched and thus coupled to modes guided in the dielectric substrate. These GMR are responsible for the narrow-line peaks and their influence on the filter's performance is thoroughly investigated.

The experimental observation of GMR at THz frequencies has been only very recently reported [38] at the frequency of approximately 7 THz. Although demonstrating the proof-of-principle, these observed GMR peaks exhibited low transmittance and broad lines owing to the use of a lossy substrate. Theoretically, it has been proposed that GMR can lead to very narrow-linewidth resonances that manifest in the spectral response of THz metamaterials [39]. In this work, we have experimentally measured THz resonances with transmittance well above 50% and full-width half-maximum (FWHM) even below 1% of the resonant frequency.

The FSS filters are fabricated using standard lithography processes, which can be scaled up to mass production using low-cost large area electronics and roll-to-roll processes. Very good agreement is observed between experimental THz-TDS measurements and finite-element numerical simulations. The resulting samples are both mechanically stable and conformable. The latter property allowed for the characterization of the filters in a bent configuration, which revealed two distinct behaviors, the robustness of the fundamental FSS filter response and the suppression of the GMR peaks. Finally, a discussion on possible applications of this novel class of FSS-THz filters in low-cost and flexible THz devices is provided.

II. NUMERICAL ANALYSIS

The schematic layout of the proposed THz-FSS filters is shown in Fig. 1. The periodic metallic FSS square lattice is characterized by the pitch P and the cross-shaped apertures are defined by the cross-arm length W and width w . The metallic screen is made of aluminum and it is supported by a thin foil of the cyclo-olefin polymer Zeonor of thickness d . The Al film thickness is 200 nm, which is thicker than the Al skin depth in the investigated frequency range, namely 150 and 58 nm at

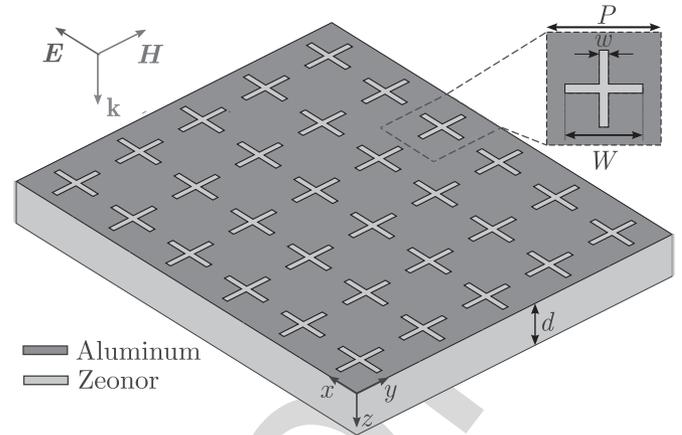


Fig. 1. Schematic layout of the investigated FSS terahertz filters. The square lattice has a pitch P and the length and width of the cross-shaped apertures are W and w , respectively. The FSS structure is patterned on a 200-nm-thick aluminum layer deposited on a Zeonor foil of subwavelength thickness d .

0.3 and 2 THz, respectively. At the same time, it is thin enough in order to avoid unnecessary stressing of the polymer during the fabrication process. The thickness of the Zeonor layer is in principle an independent variable, however in this study we focus on three values, $d = 40, 100, \text{ and } 188 \mu\text{m}$, which correspond to available films for fabrication, as it will be discussed in Section III.

The numerical simulations of the THz filters was conducted via the frequency-domain finite element method (FEM), which was implemented in the commercial software COMSOL Multiphysics. A unit cell of the periodic array was simulated by properly imposing periodic boundary conditions at the $x-z$ and $y-z$ planes. The structure was excited with an x -polarized plane wave propagating along the z -axis as in Fig. 1. The transmittance of the zero-order diffracted mode, i.e. the excited planewave, was measured at the exit of the filter, below the polymer film substrate, and normalized to the power carried by the excitation planewave. Aluminum was modeled as a Drude medium [40] and Zeonor as a dielectric with a refractive index equal to $n_z = 1.52 - j0.001$, as it has been demonstrated that it exhibits very low dispersion in the frequency range under investigation [41]–[43]. This polymeric material was selected for its excellent THz properties, namely very low-losses, high mechanical flexibility, heat resistance, and negligible birefringence. In fact, it was observed via numerical simulations that, in the context of the proposed THz-FSS filters and for the substrate thicknesses here reported, the effect of Zeonor's dielectric losses was negligible.

Fig. 2(a) provides a reference result on the transmittance of a free-standing THz filter, i.e. in the absence of the polymer substrate, for $P = 160 \mu\text{m}$, $W = 110 \mu\text{m}$, and $w = 10 \mu\text{m}$. A Lorentzian-shaped filter is observed with peak transmittance $T = 0.88$, resonant frequency $f_0 = 1.293 \text{ THz}$, and a FWHM of 200 GHz, i.e. $\sim 15\%$ of f_0 . The filtering effect stems from the response of THz wave transmission through the cross-shaped apertures owing to the resonance of the fundamental mode in the cross arm slots, which is maximized at the frequency f_0 ,

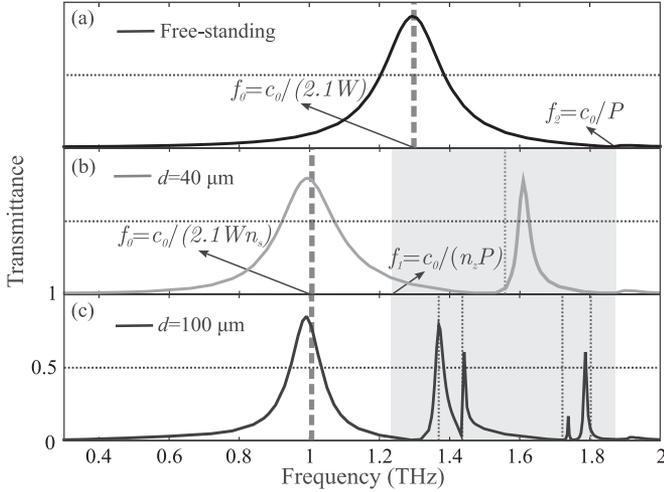


Fig. 2. (a) Power transmittance of the zero-order diffracted mode, numerically calculated for a free-standing FSS filter with $P = 160 \mu\text{m}$, $W = 110 \mu\text{m}$, and $w = 10 \mu\text{m}$. The dashed line corresponds to the frequency f_0 predicted by the approximative formula for resonant dipoles, whereas f_2 denotes the onset of the first diffractive order, associated with Wood's anomaly that leads to zero transmittance at the frequency f_2 . (b,c) Transmittance of the filter for a substrate thickness $d = 40 \mu\text{m}$ and $d = 100 \mu\text{m}$, respectively, where other parameters as in (a). The shaded region denotes the frequency interval where guided-mode resonance can manifest and the dotted lines mark the resonant frequencies predicted by GMR theory.

169 due to impedance matching [44]. When $w/(P - W) < 1$, the
 170 resonant frequency can be well approximated by the formula for
 171 resonant dipoles $f_0 = c_0/(2.1W)$ [16], marked as a dashed line
 172 in Fig. 2 and thereafter. As in other similar filters reported in the
 173 literature, this resonance only marginally depends on the pitch
 174 P , since it is not diffractive in nature, and in this work we denote
 175 it as FSS resonance (FSSR). At the frequency $f_2 = c_0/P$ that
 176 marks the threshold at which the first diffractive order appears,
 177 c_0 being the speed of light in free-space, the Wood's anomaly
 178 associated with zero transmittance is observed [21], [45].

179 Fig. 2(b) and (c) investigate the same FSS structure, albeit in
 180 the presence of the Zeonor substrate, with a thickness of 40 and
 181 100 μm , respectively. Compared to the free-standing reference
 182 case of Fig. 2(a), two major differences are observed. First, the
 183 FSSR is shifted towards lower frequencies by a factor

$$n_s = \sqrt{\frac{(n_{zr}^2 + 1)}{2}}, \quad (1)$$

184 where n_{zr} is the real part of n_z , and n_s is the index correspond-
 185 ing to the average permittivity of materials on the two sides
 186 of the FSS, namely air and Zeonor. Equation (1) is valid for
 187 substrate thicknesses higher than one tenth of the THz wave-
 188 length [21], which is the case for the considered values of d .
 189 In general, in the presence of a substrate or superstratum the
 190 THz wave has a shorter wavelength in the dielectric medium,
 191 the cross dimensions become electromagnetically larger and
 192 the FSSR frequency decreases. Second, apart from the FSSR,
 193 other resonances are observed, whose number and position de-
 194 pends on the polymer thickness. These resonances stem from
 195 the coupling of waves diffracted on the periodic FSS screen to

196 propagating modes in the substrate, which can be thought of
 197 as a dielectric slab waveguide, a phenomenon known as GMR.
 198 Grating filters based in GMR have been long known in the field
 199 of optics and photonics [46] and recently it has been shown that
 200 bandpass GMR filters can also be designed at THz frequencies,
 201 where the role of the anti-reflecting surface [47] can be played
 202 by the metallic FSS layer [38].

203 According to GMR theory, first-order resonances for normal
 204 incidence can be observed in the interval

$$\frac{c_0}{n_{zr}P} \leq f_r \leq \frac{c_0}{P}, \quad (2)$$

205 at those resonant frequencies f_r that satisfy $n_i = c_0/(f_r P)$,
 206 where n_i is the effective index of a mode guided in the substrate
 207 slab waveguide. In Fig. 2(b) and (c) we have annotated with
 208 grey shading the spectral window where GMR can occur. In-
 209 side these regions, we have calculated a set of GMR frequencies,
 210 marked as dotted lines, according to the following steps: first, the
 211 resonant frequencies f_r calculated by the FEM simulations are
 212 identified, i.e. the transmission maxima in the gray-shaded areas
 213 of the calculated spectra. Then, for each slab thickness the effec-
 214 tive modal indices at f_r are calculated using a freely available
 215 electromagnetic mode solver for 1-D dielectric multilayer slab
 216 waveguides [48]. Among the resulting modal indices $n_i(f_r)$, the
 217 frequencies $c_0/(n_i P)$ are calculated and the one closely match-
 218 ing f_r is marked, with each resonant frequency associated with
 219 a different slab mode. Better agreement is achieved for higher
 220 values of d and for modes closer to the limit $f_1 = c_0/(n_{zr}P)$. In
 221 both cases, these modes show higher confinement thanks to ei-
 222 ther the higher slab thickness or the higher effective modal index
 223 and hence lower modal order. The discrepancy between GMR
 224 theory and FEM simulations is attributed to the presence of the
 225 reflecting metallic FSS screen, which introduces a perturbation
 226 in the geometry of the slab waveguide.

227 It is clear that the positions of the GMR frequencies depend
 228 strongly on d and P , as these parameters control the matching
 229 condition between the wave vector of the diffracted orders and
 230 the propagating substrate modes. This strong dependence is not
 231 to be expected as far as the exact geometry of the cross-shaped
 232 apertures is concerned. In order to further elucidate this point,
 233 we have calculated the transmittance of a series of filters with
 234 fixed $d = 100 \mu\text{m}$, $P = 160 \mu\text{m}$, $w = 10 \mu\text{m}$, for various val-
 235 ues of the cross-arm length W . The results reported in Fig. 3
 236 demonstrate that the GMR are only slightly affected by the
 237 variation of W , as summarized in the results of Table I, while
 238 the opposite stands for the FSSR, which depends on W via
 239 $f_0 = c_0/(2.1Wn_s)$. Fig. 4 investigates a complementary sce-
 240 nario, namely the variation of the pitch P for an FSS with fixed
 241 $W = 110 \mu\text{m}$, $w = 10 \mu\text{m}$, for the three available thicknesses
 242 of the Zeonor film thickness. In this case it is the FSS resonant
 243 frequency f_0 that remains approximately at the same position,
 244 while those of the GMR shift towards lower frequencies for
 245 higher pitch values, given the condition described by (2). More-
 246 over, as d gets higher, a larger number of GMR is supported
 247 that corresponds to a higher number of modes guided in the slab
 248 waveguide, whose positions are well resolved by the numerical
 249 simulations described above.

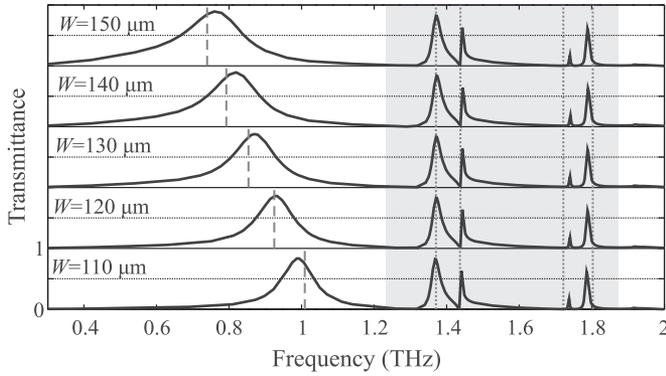


Fig. 3. Transmittance of the FSS filter for $d = 100 \mu\text{m}$, $P = 160 \mu\text{m}$, $w = 10 \mu\text{m}$, for various values of the cross-arm length W . The main resonant frequency f_0 shifts towards lower frequencies for higher W , whereas the GMRs remain unaffected. The spectral position of the GMRs is well approximated by calculating the frequencies $f_i = c_0 / (n_{\text{eff},i} P)$, marked as dashed lines, where n_i are the effective indices of the modes supported by the single Zeonor slab waveguide.

TABLE I
SIMULATED GUIDED-MODE RESONANT FREQUENCIES (IN THZ)
FOR THE FILTERS STUDIED IN FIG. 3

W (μm)	$f_r^{1,\text{sim}}$	$f_r^{2,\text{sim}}$	$f_r^{3,\text{sim}}$	$f_r^{4,\text{sim}}$
150	1.371	1.443	1.743	1.797
140	1.372	1.445	1.749	1.803
130	1.373	1.448	1.753	1.809
120	1.375	1.451	1.753	1.815
110	1.379	1.453	1.753	1.817

The resonances predicted via GMR theory (dashed lines) occur at $f_r^1 = 1.37 \text{ THz}$, $f_r^2 = 1.438 \text{ THz}$, $f_r^3 = 1.733 \text{ THz}$, and $f_r^4 = 1.803 \text{ THz}$, and the corresponding slab modal indices [48] are $n_1 = 1.368$, $n_2 = 1.302$, $n_3 = 1.081$, and $n_4 = 1.038$.

250 Apart from the interesting underlying physics, the GMR fil-
251 ters show also great potential in view of THz applications that
252 need narrow-line selective filtering. In this respect, one main
253 drawback of the traditional FSSR filters is that their linewidth,
254 defined as FWHM/f_0 , is in the range $5\% \sim 20\%$, according to
255 the various designs [38]. Achieving more narrow filters is pos-
256 sible by reducing the aperture's dimensions, particularly w , or
257 by stacking more than one FSS filters, although this comes to
258 the expense of significantly reduced transmittance. On the con-
259 trary GMR-based THz filters can have linewidths lower than 1%
260 [38], without compromising the filter's transmittance, which is
261 validated both numerically and experimentally in this work.

262 III. EXPERIMENTAL DEMONSTRATION

263 We have experimentally investigated the properties of the
264 proposed THz filters by fabricating samples with different pe-
265 riod and cross-arm length values on low-loss flexible 40, 100,
266 and $188\text{-}\mu\text{m}$ -thick Zeonor foils using standard photolithogra-
267 phy techniques. First, an aluminum layer of 200 nm thickness
268 was thermally evaporated on Zeonor foils. Subsequently, a film
269 of the positive photoresist S1813 from Shipley was deposited

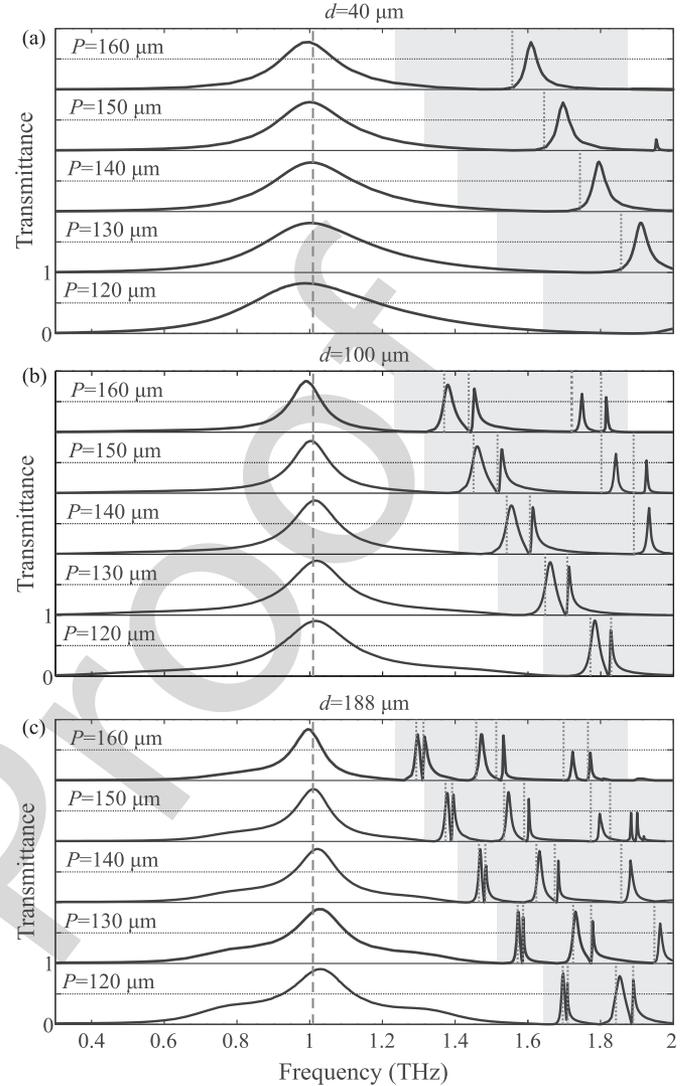


Fig. 4. Transmittance of the FSS filter numerically calculated for $W = 110 \mu\text{m}$, $w = 10 \mu\text{m}$, various values of the lattice pitch P and for three commercially available thicknesses of the Zeonor substrate, namely (a) $d = 40 \mu\text{m}$, (b) $d = 100 \mu\text{m}$, and (c) $d = 188 \mu\text{m}$. The main resonant frequency f_0 remains fixed at $f_0 \approx c_0 / (2.1 P n_s)$, whereas the GMRs shift as a function of the lattice pitch and the substrate thickness.

270 by spin-coating at 3000 rpm for 30 seconds and then cured at
271 $115 \text{ }^\circ\text{C}$ for 2 minutes . The resulting thickness of the photoresist
272 layer was $1.3 \pm 0.1 \mu\text{m}$. Photolithography was carried out on
273 the metalized surface using a Karl Suss MA150 mask aligner
274 with a wavelength of 365 nm and intensity of $60 \text{ mW}/\text{cm}^2$. The
275 samples were immersed in the developer MF319 for 50 seconds ,
276 rinsed with deionized water, dried with nitrogen and cured at
277 $120 \text{ }^\circ\text{C}$ for 5 minutes . Then, the exposed aluminum was wet-
278 etched in $\text{H}_3 \text{ PO}_4 : \text{H}_2 \text{ O} : \text{CH}_3 \text{ COOH} : \text{HNO}_3 = 16:2:1:1$ and the
279 residual photoresist was removed with acetone. The filters were
280 cut in samples of $2 \text{ cm} \times 2 \text{ cm}$.

281 The transmission properties of the fabricated THz filters were
282 investigated by means of THz time domain spectroscopy using
283 a Menlo Systems TERA K15 THz-TDS all fiber-coupled spec-
284 trometer in transmission mode using collimated and polarized

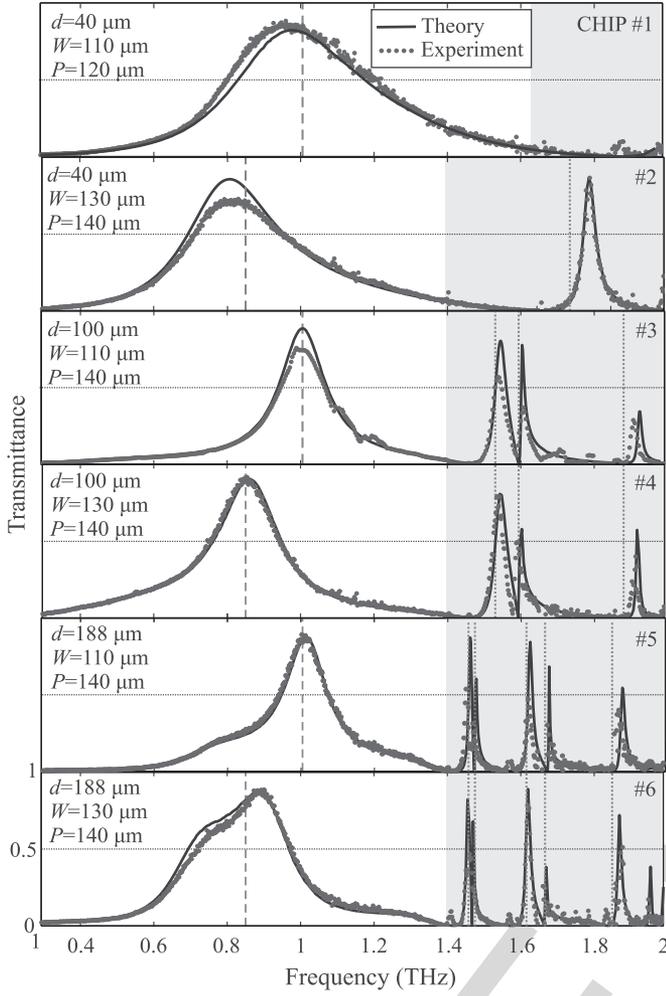


Fig. 5. Direct comparison of the FSS filter's transmittance between numerical FEM simulations and experimental TDS measurements for a series of fabricated samples with different geometrical parameters.

radiation. The power transmittance measured for each sample was normalized to that of the reference signal, i.e. in the absence of the sample. The spot-size of the collimated beam was approximately 10 mm in diameter and a time scan of 400 ps was employed for a spectral resolution of 2.5 GHz. The measurements were done in an atmosphere purged with nitrogen to prevent the absorption of THz radiation from water vapor in the air.

Fig. 5 shows a direct comparison between the numerically simulated transmittance of the THz filters, calculated by means of the finite-element method, and the experimental TDS measurements for six different chips, two for each one of the available Zeonor foil substrates. The experimental results reproduce very well the numerical simulations, in terms of both the position and the lineshape of the various transmission peaks. For $d = 40 \mu\text{m}$ and $P = 140 \mu\text{m}$ (Chip #2) two clearly separated resonances are observed, the FSSR at 0.8 THz and a single GMR at 1.8 THz. It is experimentally verified that, owing to the different underlying physical mechanisms, these two resonances show very different linewidths: the FWHM measured for

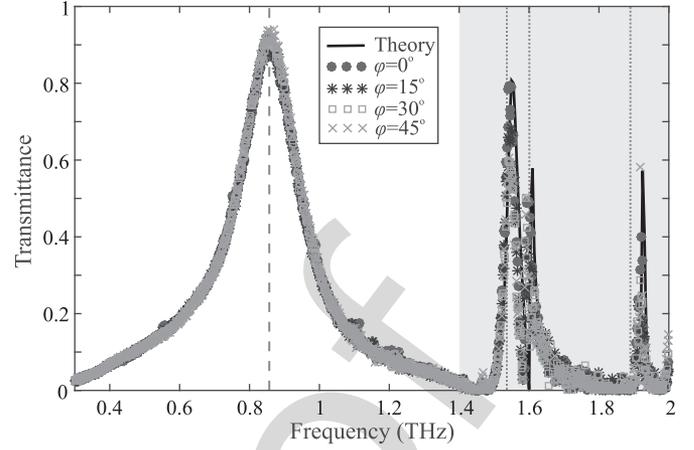


Fig. 6. TDS measurements of Chip #4, as in Fig. 5, for different angles φ of the sample's rotation in the $x - y$ plane, demonstrating polarization-insensitive operation.

the FSSR and GMR is 40% and only 2.3% of the resonant frequency, respectively. For $d = 100 \mu\text{m}$ and $188 \mu\text{m}$, the existence of closely spaced GMR peaks, stemming from the excitation of more modes in the slab waveguide substrate, leads to asymmetric Fano-like linewidths [49], with high transmittance and even more narrow linewidths. For instance, Chip #4 is characterized by three GMR at 1.546, 1.597, 1.921 THz with linewidths FWHM/ f_0 equal to 2.5%, 1.3%, and 1%, respectively, while the high-transmittance resonances at 1.63 and 1.88 THz for Chip #6 exhibit corresponding linewidths of 1% and 0.7%.

It is observed that in the case of some GMR the experimentally measured transmittance peak values are somewhat lower than the numerically calculated prediction. This is attributed to three factors: a) the very narrow linewidths of such resonances, particularly for $d = 188 \mu\text{m}$, which are comparable with the TDS resolution, b) minor defects in fabrication or the planarity of the samples, c) non-ideal collimation/ residual divergence of the spot, and d) the finite dimensions of the sample and THz spot, with a diameter of a few tens of wavelength in size, in contrast to the infinite periodic array assumed in the simulations. The latter, in particular, is very relevant for GMR, since these are numerically simulated as the result of constructive interference of the excited waveguide modes and the diffractive waves on an infinite periodic FSS metallic screen, while the measurements are conducted over a finite truncated lattice. Nevertheless, apart from these small discrepancies, it is overall demonstrated that the fabricated FSS-THz filter can achieve both broad- and narrow-band filtering, depending on the selection of the geometrical parameters.

An important trait of the proposed THz filters is that they are polarization-independent, owing to the square FSS lattice and the symmetry of the cross-shaped apertures. This has been experimentally verified by rotating the fabricated samples in the $x - y$ plane, i.e. perpendicular to the propagation direction of the x -polarized THz wave, and recording the measured TDS spectra. Fig. 6 shows a set of results obtained for the filter characterized by $P = 140 \mu\text{m}$, $W = 110 \mu\text{m}$, $w = 10 \mu\text{m}$,

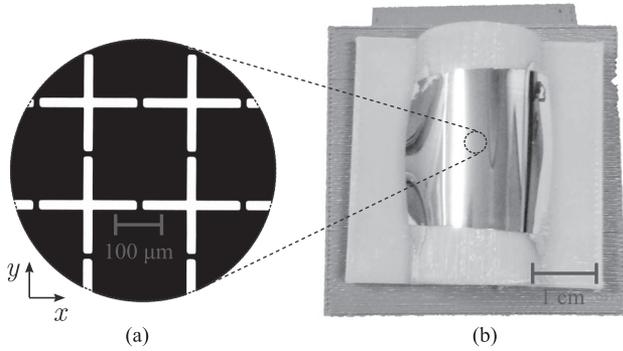


Fig. 7. (a) Micrograph taken under the microscope in transmission mode of a fabricated sample with $P = 140 \mu\text{m}$, $W = 130 \mu\text{m}$, and $w = 10 \mu\text{m}$ on a Zeonor foil with $d = 100 \mu\text{m}$. (b) The same sample bent at a radius of 1 cm and fixed on a properly assembled mount in order to characterize its properties as a conformal THz filter.

and $d = 100 \mu\text{m}$ (Chip #4 with reference to Fig. 5), where the angle φ denotes the rotation angle of the sample, measured from the x -axis. The spectra recorded for $\varphi = 0^\circ$, 15° , 30° , and 45° overlap, thus demonstrating the polarization-independent response of the THz filter.

Among the appealing properties of the employed thin polymer films are their flexibility and ability to easily conform to curved surfaces [6]. In this work, we have experimentally investigated the transmission properties of Chip #4, when bent down to a curvature with radius 1 cm. The inset of Fig. 7(a) shows a micrograph of the fabricated filter taken under optical microscope in transmission mode with a 20x microscope objective, where the black parts are aluminum, while the transparent foil appears white. Fig. 7(b) shows the bent sample mounted on a properly designed frame, so that it is placed in the THz beam path of the TDS setup. It is remarked that, after processing, the filter does not show any buckling and maintains its mechanical and electromagnetic properties after bending it several times.

Fig. 8 shows the experimental characterization of the bent THz filter, where the numerical and experimental results for the flat configuration are also reported for comparison. In the experimental characterization, the incoming THz wave was polarized along the x -axis, as defined in Fig. 7. It is evident that the bent filter retains its filtering property as far as the FSSR is concerned, while all remaining peaks that stem from GMR excitation are no longer observed. These interesting features can be explained by taking into account the physical origin of the filter's resonances. The FSSR involves the excitation of a localized mode inside the cross-shaped aperture, which depends on the aperture's dimensions and is not diffractive in nature, hence the very weak dependence of its central frequency on the lattice pitch, as demonstrated in Fig. 4. Also, the presence of the substrate induces a shift of the resonant frequency by the factor n_s , but does not otherwise affect the transmission mechanism of the filter.

On the contrary, the GMR are excited due to the coupling of first-order diffracted waves into the substrate slab modes. This coupling is strongly dependent on both the lattice pitch and the polymer film's thickness. When the sample is bent,

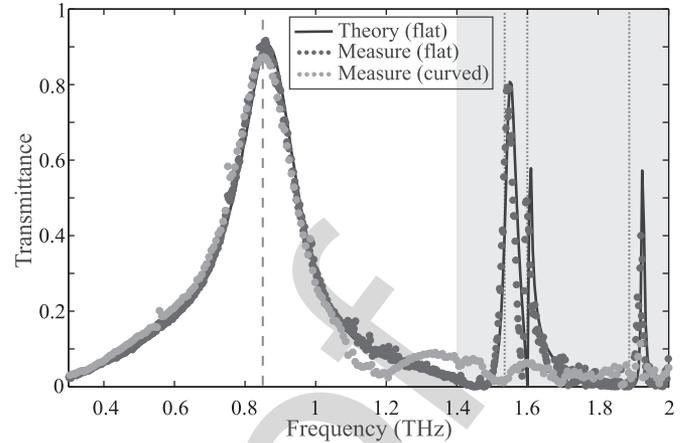


Fig. 8. Comparison of the experimentally measured transmittance of the FSS filter shown in Fig. 7 between the flat and bent configuration. The numerical simulations for the flat case are also shown for reference. The main FSS resonance remains unaffected, demonstrating the capability of the fabricated samples to operate as flexible and conformal thin film THz filters. The GMR peaks in the bent configuration are suppressed.

impinging THz plane wave does not sample the same effective pitch across the filter's surface and the substrate is no longer a flat slab dielectric waveguide. This leads to the suppression of GMR peaks, as evidenced in the results of Fig. 8. Therefore, apart from the filtering properties already demonstrated in Fig. 5 for the flat configuration, the proposed THz filters show very interesting properties also when bent, which can be readily exploited in various applications. For instance, flexible filters based on the FSSR can be designed for use as thin conformal layers on curved surfaces or components in THz setups. Moreover, the sensitivity of the GMR transmittance and central frequencies on the bending radius can provide the basis for the development of sensor devices, for instance curvature sensors or components for the measurement of the thickness and/or refractive index of thin dielectric layers at THz frequencies by placing the FSS-GMR filters on top of the sample and measuring the shift of the GMR frequencies.

IV. CONCLUSION

In brief, we have numerically and experimentally investigated a new class of THz filters based on the patterning of metallic cross-shaped FSS on thin films of the low-loss cyclo-olefin polymer Zeonor. By properly adjusting the geometrical parameters of the device both broad- and narrowline filters can be designed, the first stemming from the transmittance of THz waves through FSS cross-shaped apertures, while the latter from the excitation of guided-mode resonances in the polymer substrate. Not observed before in this kind of FSS structures, the GMR filters show extremely narrow linewidths with high transmittance. The FSS filters are shown robust to the bending of the flexible Zeonor films, thus paving the way for conformal THz filters integrated on curved surfaces. On the contrary, the diffractive nature of GMR renders them sensitive to deformations or changes of

substrate's properties, properties that could be exploited in the design of sensors working at THz frequencies.

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